

FOURIERTRANSFORMEN

- ◆ Fouriertransformen till $x(t)$:

$$\mathcal{F}\{x(t)\} = X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$\left(\begin{array}{l} \text{Jfr. fourierserie:} \\ C_k = \frac{1}{T} \int_{\alpha}^{\alpha+T} x(t) e^{-jk\omega_1 t} dt \end{array} \right)$$

- ◆ Inversa fouriertransformen till $X(\omega)$:

$$\mathcal{F}^{-1}\{X(\omega)\} = x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

$$\left(\begin{array}{l} \text{Jfr. fourierserie:} \\ x(t) = \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_1 t} \end{array} \right)$$

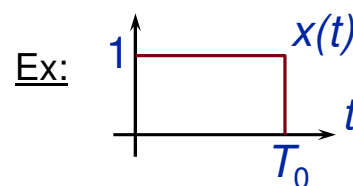
- Existensvillkor:

$$\mathcal{F}\{x(t)\} \exists \text{ om } \int_{-\infty}^{\infty} |x(t)| dt < \infty$$

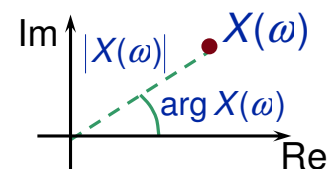
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Frekvensgenskap hos signal

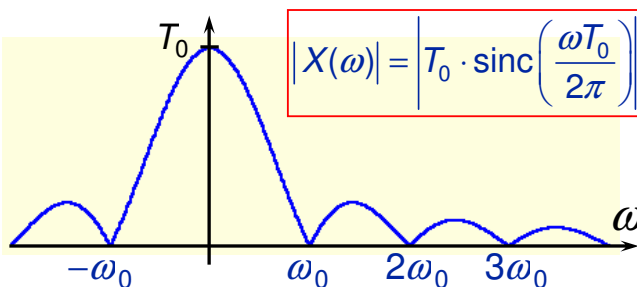
- ◆ Frekvensspektrum, $X(\omega)$:



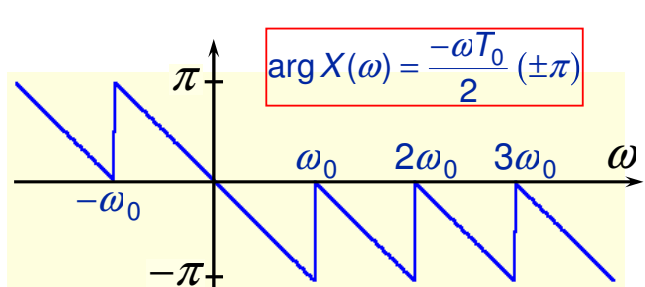
$$X(\omega) = e^{-j\frac{\omega T_0}{2}} \cdot T_0 \operatorname{sinc}\left(\frac{\omega T_0}{2\pi}\right) = |X(\omega)| e^{j\arg X(\omega)}$$



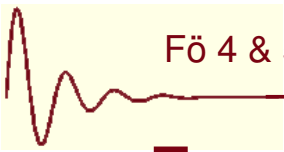
Amplitudspektrum, $|X(\omega)|$:



Fasspektrum, $\arg X(\omega)$:



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Fouriertransform till distribution

- ♦ Utöka klassen av fouriertransformerbara funktioner (fouriertransform till begränsad, ej absolutintegrerbar signal):

Låt $g(t)$ utgöra en snabbt avtagande (och mycket snäll) fouriertransformerbar testfunktion.

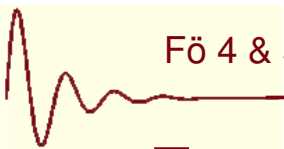
Den *distribution* X som då uppfyller sambandet

$$\int_{-\infty}^{\infty} X(\lambda)g(\lambda) d\lambda = \int_{-\infty}^{\infty} x(\lambda)G(\lambda) d\lambda$$

definieras som fouriertransformen till *distributionen* x .

(även $G(\lambda) = \mathcal{F}\{g(t)\}$ är en snabbt avtagande och mycket snäll testfunktion)

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Energispektrum

- ♦ Låt $x(t)$ vara en reellvärd spänning (eller ström) som läggs över (går genom) en resistans på 1Ω .

Energiinnehållet i $x(t)$ är då $W = \int_{-\infty}^{\infty} x^2(t) dt$

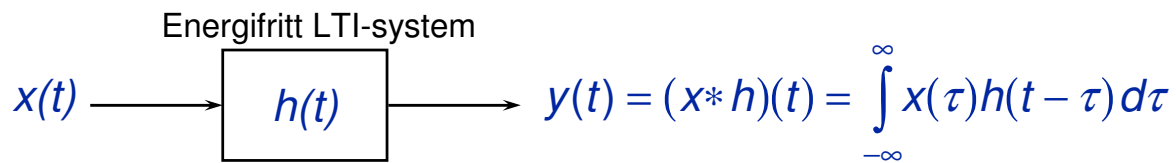
- ♦ Parsevals formel gäller generellt för komplexvärd fouriertransformerbar signal $x(t)$:

Signalenergin $W = \int_{-\infty}^{\infty} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(\omega)|^2 d\omega$

$|X(\omega)|^2$: Energispektrum

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SYSTEMANALYS



$$\underline{Y(\omega)} = \mathcal{F}\{y(t)\} = \dots \dots \dots = \underline{X(\omega)H(\omega)}$$

$$\Rightarrow \text{Faltningsteoremet: } \mathcal{F}\{f_1 * f_2\} = F_1(\omega)F_2(\omega)$$

- ◆ Frekvensfunktion: $H(\omega) = \mathcal{F}\{h(t)\} = |H(\omega)| e^{j\arg H(\omega)}$
 - Amplitudkaraktäristik: $|H(\omega)|$
 - Faskaraktäristik: $\arg H(\omega)$

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Systemanalys, forts.

$$Y(\omega) = X(\omega)H(\omega) \Rightarrow \begin{cases} |Y(\omega)| = |X(\omega)| \cdot |H(\omega)| \\ \arg Y(\omega) = \arg X(\omega) + \arg H(\omega) \end{cases}$$

$$\Rightarrow |Y(\omega)|^2 = |X(\omega)|^2 \cdot |H(\omega)|^2$$

Energiöverföringsfunktion

- ◆ Stationär sinus:

Insignal: $\underline{x(t)} = \hat{X} \sin(\omega_0 t + \varphi) = \text{Im}\{\hat{X} e^{j(\omega_0 t + \varphi)}\}$

$$\begin{aligned} \Rightarrow \underline{y(t)} &= (x * h)(t) = \dots = \text{Im}\{\hat{X} e^{j(\omega_0 t + \varphi)} \cdot H(\omega_0)\} \\ &= \underline{\hat{X} \cdot |H(\omega_0)| \sin(\omega_0 t + \varphi + \arg H(\omega_0))} \end{aligned}$$

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Kretsberäkningar, linjära *RLMC*-nät

(passiva kretselement, fouriertransformerbara källor)

METODIK, beräkna godtycklig nätspänning / -ström:



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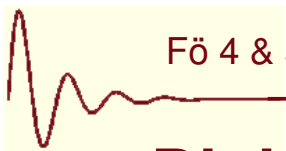
Kretsberäkningar, metodik (forts)



4) Likströmsteori \Rightarrow Sökt storhets fouriertransform ($Y(\omega)$)

5) Inverstransformera \Rightarrow Sökt storhets tidsuttryck ($y(t) = \mathcal{F}^{-1}\{ Y(\omega) \}$)

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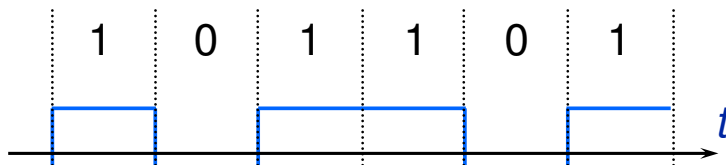


Digital kommunikation

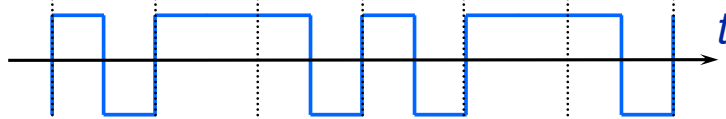
Digital signalering med analoga signalvågformer

Basbandsmodulation,

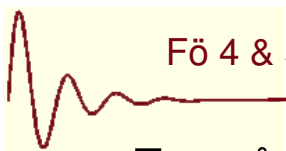
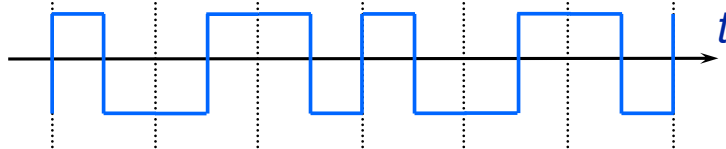
Exempel 1:



Exempel 2:

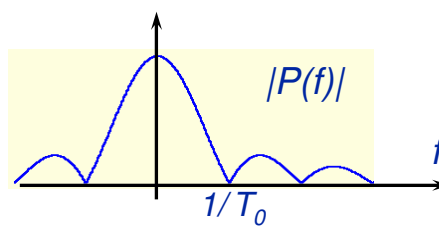
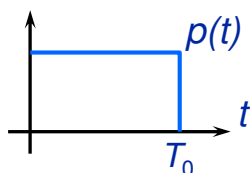


Exempel 3:

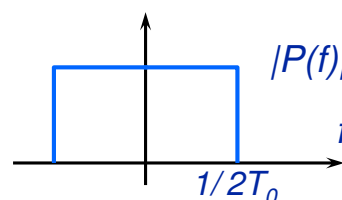
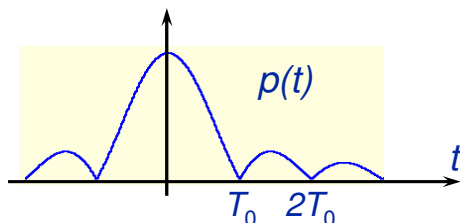


Ex. på signalpulsformer för basbandskanaler:

$$p(t) = u(t) - u(t - T_0)$$

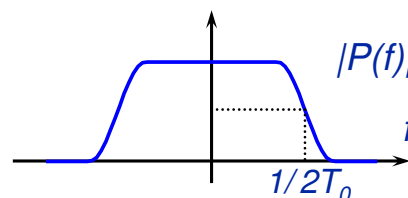
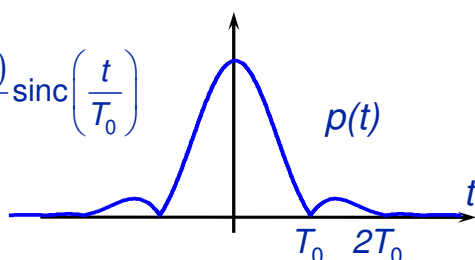


$$p(t) = \text{sinc}\left(\frac{t}{T_0}\right)$$



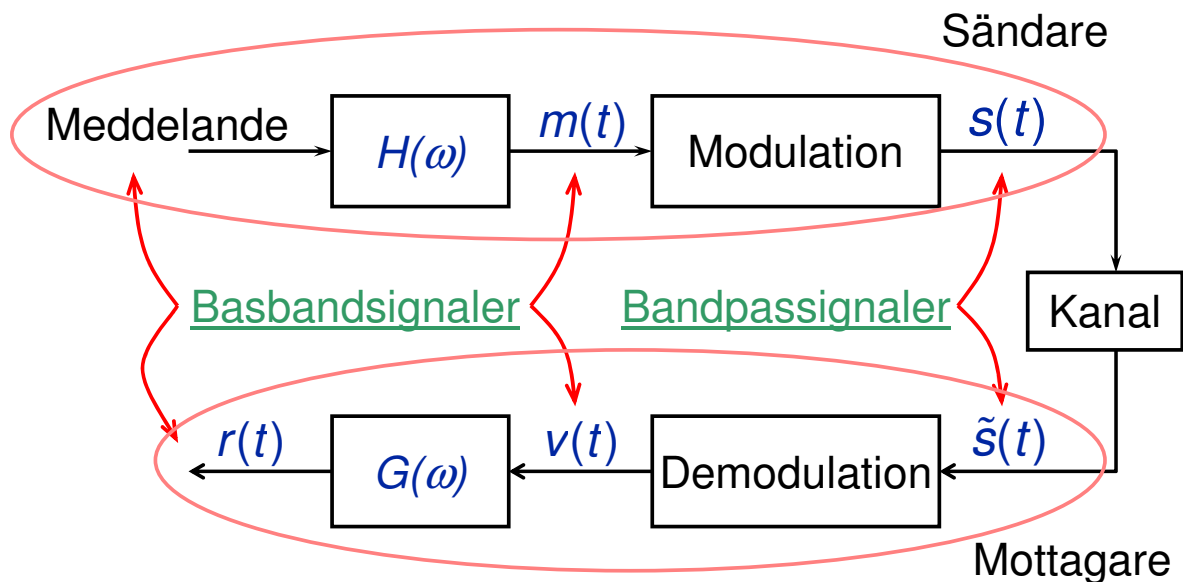
$$p(t) = \frac{\cos(2\beta\pi t/T_0)}{1 - (4\beta t/T_0)^2} \text{sinc}\left(\frac{t}{T_0}\right)$$

"Raised cosine"



Vanligt: högfrekvent signalering (Ex: ADSL, radio- och satellitkommunikation, m.m.)

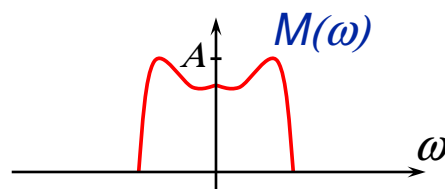
- ◆ Typiskt analogt kommunikationssystem:



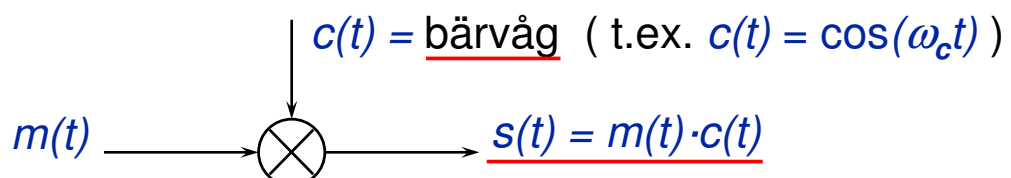
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Generell Amplitudmodulering

- ◆ Basbandsignal (här: meddelandesignalen $m(t)$):



- ◆ (Amplitud-)Modulering:

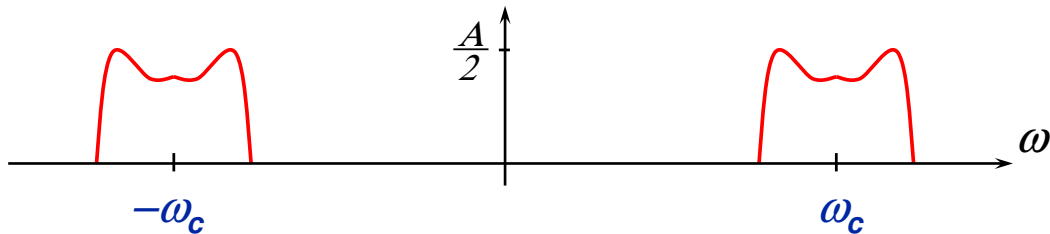


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Amplitudmodulering, forts

- ◆ Bandpassignal:

$$\underline{S(\omega) = \mathcal{F}\{m(t) \cdot c(t)\} = \frac{1}{2\pi}(M * C)(\omega)}$$



där $C(\omega) = \mathcal{F}\{\cos(\omega_c t)\} = \pi(\delta(\omega + \omega_c) + \delta(\omega - \omega_c))$

$$\Rightarrow S(\omega) = \frac{1}{2}(M(\omega + \omega_c) + M(\omega - \omega_c))$$

Amplitudmodulering, forts

- ◆ Demodulering + LP-filter:

